A CONTACTLESS ELECTROCHEMICAL IMPEDANCE MEASUREMENT METHOD

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ABSTRACT

We present the first demonstration of a passive, contactless method to interrogate electrochemical impedance (EI) via a microfabricated MEMS coil. EI measurement was achieved using the principle of reflected impedance between a pair of inductively coupled coils. Ionic concentration and electrolyte temperature were wirelessly transduced from measurement electrodes connected to a microfabricated coil in a flexible thin film substrate. This technique is widely applicable for resistive transduction and obviates the need for soldering discrete electronic components, microcontrollers, batteries, and bulky hermetic packaging. It also features low power consumption (~75 μ W) and all biocompatible construction (only Au, Ti, and Parylene C), making it ideal for use in harsh or *in vivo* environments.

INTRODUCTION

Chronic recording of physiological signals via the use of implanted sensors has been a long-standing goal for the medical and scientific community. Such sensors would enable continuous monitoring to facilitate greater understanding of disease progression as well as inform course of treatment.

The *in vivo* environment poses a formidable challenge for the development of telemetry circuitry used to encode and wirelessly transmit measured signals [1]. Water vapor and ion permeation create undesired paths for electric current and can alter critical bias voltages. The inflammatory response to foreign bodies can subject the implant to a highly oxidative or acidic environment [2]. Blood flow and other involuntary movements can subject the implant to chronic and repeated mechanical strain. Such events can damage sensitive components or introduce errors in signal measurement and transmission.

Naturally, the presence of the implant should also minimize harm to the host environment. The materials selected for the packaged device should be nontoxic and induce minimal immune response. The use of soft, flexible materials (i.e. those mechanically matching tissue properties) and elimination of sharp corners also serves to minimize damage to tissue [3]. Power requirements for electronics should be kept in check to minimize heating to safe levels.

In this work, our goal is to address these requirements with a new method to measure and transmit signals within a simulated *in vivo* environment. Passive wireless capacitive transduction methods have been widely studied and utilized, with notable examples demonstrating successful chronic implantation in humans [4]. However, for resistive transduction, wireless transmission has been limited to the use of active circuitry (analog or digital via microcontrollers). This burdens system design by adding power constraints, thermal budget concerns, and bulk to the overall package. Here, we introduce the use of reflected impedance across a pair of inductively coupled coils to achieve wireless resistive transduction without active circuitry and demonstrated its use for measurement of electrolyte conductivity and temperature.

THEORETICAL CONSIDERATIONS

A pair of inductively coupled coils shares the properties of a transformer. An alternating current source induces a varying

magnetic field in the primary coil, which then induces a varying electromotive force in the secondary coil (Figure 1).



Figure 1: Schematic of reflected impedance principle, where Z_p represents the complex impedance of the primary circuit resistance and capacitance, L_p and L_s the inductances of coupled coils, and Z_s the complex impedance of the load on the secondary circuit. An AC source (LCR meter) induces current within the secondary circuit across a pair of inductively coupled coils (coupling coefficient M).

To simplify, an equivalent single loop model that lumps the secondary circuit into a complex impedance term may be utilized (Figure 2). Due to conservation of energy, a load that is attached to the secondary coil will result in an apparent load on the primary side, as represented in (1). In this manner, changes in the secondary load may be reflected across the inductively coupled coils and measured as changes on the primary side [5].



Figure 2: Equivalent single loop primary circuit with reflected impedance represented as Z_R .

$$Z_R = \frac{\omega^2 M^2}{Z_s + j\omega L_s} \tag{1}$$

DESIGN

Size of the secondary coil was maximized to optimize coupling with the primary coil. Coupling can also be improved by increasing the number of turns in order to capture increased magnetic flux. However, this can result in a large and unwieldy device footprint for implant applications. Additional improvement in coupling can be achieved through the use of a ferrite core on the primary coil.

All of the materials selected for use have a proven track record for use within the *in vivo* environment. Thin film gold was used for the electrodes due to its inertness and its well-characterized electrochemical properties [6]. Electrode size and spacing were based on previous investigations on electrochemical impedance sensing [7]. The low resistivity of gold also serves advantageous for its use as a coil in RF applications. Parylene C was selected as the substrate and to encapsulate the electrodes and coil, due to its excellent insulation properties as well as its amenability to microfabrication processes [8].

METHODS

The fabrication of this thin film polymer device was based on previously reported methods [7]. A pair of Au electrodes (2000 Å thick on 200 Å Ti adhesion layer, $300 \ \mu m \times 300 \ \mu m$, $500 \ \mu m - 5000 \ \mu m$ electrode-to-electrode spacing) attached to a planar coil (30 mm diameter, 1-16 turns) was patterned and electron beam evaporated onto a flexible Parylene C substrate (10 \mu m thick) (Figure 3). Following deposition of another 10 \mu m Parylene insulation layer, electrodes were exposed via O₂ plasma reactive ion etching (Technics RIE, 200 W, 150 mTorr) [9].



Figure 3: Overview of fabrication process. (i) Deposition of 10 μ m thick Parylene C substrate. (ii) Patterning and deposition of Ti adhesion layer and Au. (iii) Deposition of 10 μ m Parylene insulation layer, reactive ion etching (O₂) to expose electrode sites, release, and cutout from silicon substrate.

After cutout and release (Figure 4), the devices were placed in a polystyrene Petri dish and submerged in phosphate buffered saline (PBS) of various concentrations $(0.5 - 10 \times, 8-150 \text{ mS/cm [10]})$ and temperatures (22°C - 65°C) to mimic a range of typical and abnormal *in vivo* conditions. Deionized water was used to establish a baseline impedance reference.



Figure 4: (Left) Photograph of microfabricated and released thin film Parylene C device held by forceps. Sixteen turn coil shown; one, five, and six turn coils were also fabricated. (Right) Inset of left; micrograph highlighting electrochemical impedance (EI) measurement electrodes ($300 \times 300 \mu m$ exposed area).

The secondary circuit comprises of the Au planar coil ($L = 11.1 \mu$ H, $Q \sim 1.5$ at 2.198 MHz) and electrodes submerged in electrolyte, which can be represented by the Randles circuit model (Figure 5). At resonant frequency, the reactance of the capacitive (C_{dl}) and inductive (thin film coil) elements will be equal in magnitude but opposite in sign. The resulting reflected impedance consists of the remaining resistive element, which is the solution resistance (R_s). The Petri dish (0.8 mm thick base) was placed onto and concentrically aligned with the commercially available primary coil

having a ferrite core (Figure 6, 24 μ H, Q = 180@125 kHz, Wurth Electronics, 50 mm diameter, 22 turns).



Figure 5: Secondary circuit schematic, which comprises of insulated Au coil and Randles circuit model of electrodes submersed in electrolyte.



Figure 6: Primary coil with ferrite core to improve coupling with thin film (secondary) coil.

The primary coil was elevated (onto an additional Petri dish to prevent inadvertent coupling with the benchtop) and attached to an LCR meter (Hewlett Packard E4285A) (Figure 7), which provided the AC signal and recorded the reflected impedance.



Figure 7: Device under test (DUT) was placed within a Petri dish and submerged in PBS (phosphate buffered saline). Primary coil includes ferrite core to improve inductive coupling with thin film (secondary) coil. Not drawn to scale.

The optimal measurement frequency was determined to be 2.198 MHz (Q = 34.3, minimum phase, 1 V_{P-P}) to bypass the double layer capacitance (C_{dl}) in the Randles circuit (Figure 8) [11]. At this measurement frequency, the impedance is dominated by the solution resistance. Reproducibility of coil position was achieved by observing phase angle measurement and adjusting alignment until minimum phase ($< \pm 1^{\circ}$) was attained. To assess misalignment performance, coil centers were translated up to 5 mm from initial alignment while measuring reflected impedance.



Figure 8: Frequency response of system comprising of the primary coil coupled with the DUT. The excitation signal on the LCR meter was set to 2.198 MHz (minimum system phase to bypass double layer capacitance C_{dl}) for subsequent testing.

RESULTS

Reflected electrochemical impedance was measured from fabricated devices and utilized to transduce ionic concentration and temperature of the electrolyte into which the devices were submerged (Figure 9). The observed power draw at the primary coil was approximately 75 μ W.



Figure 9: Reflected impedance was observed to be inversely proportional to ionic concentration. Deionized water was used as a baseline reference and indicated using the arrow.

Electrochemical impedance of ionic solutions has previously been shown with a direct wired measurement to be inversely proportional to temperature [12]. This relation was measured wirelessly using this reflected impedance method with $1 \times PBS$ across a range of different temperatures (Figure 10).



Figure 10: Electrochemical impedance of $1 \times PBS$ was observed to be inversely proportional to temperature.

Reflected impedance increased linearly with coil misalignment by approximately 1% per mm of coil center offset. The results were achieved with a 5 turn thin film coil, and preliminary results suggest that altering coil diameter and increasing turn count can reduce misalignment effects on baseline impedance measurement.



Figure 11: Reflected impedance increased with coil misalignment across the range tested, but remained within 5% of baseline measurement.

CONCLUSION

A method to wirelessly measure electrochemical impedance was developed and demonstrated by measuring ionic conductivity (0-150 mS/cm) and solution temperature (22-65°C). The method utilizes inductively coupled coils and requires no active electronic components. Measurement frequency was determined to be 2.198 MHz and performance with suboptimal coil alignment (translation up to 5mm) was evaluated. Featuring all biocompatible material construction, flexible polymer substrate, and low power draw (<100 μ W), the device is well suited for the wet, *in vivo* environment. The overall structure may also lend itself well to use in other harsh environments.

The electrochemical impedance transduction principle utilized here was previously demonstrated in a variety of sensing applications such as sensing of catheter blockage [13], liquid flow measurement [12], contact force [7], and drug delivery volume [10, 14], but with direct wired measurements. We demonstrated that it may be feasible to incorporate a coil and adjust electrode packaging to enable wireless subcutaneous sensing of these aforementioned signals. In addition, coating the electrode surface with antigens [15] or chemically sensitive polymers [16] can add specificity to the electrochemical measurement.

Additional development is planned to refine this method for implant applications. Miniaturization of coil size and further characterization of inductive coupling (foveation, distance, increasing Q factor) will be performed. Performance will be assessed in an *ex vivo* application to measure catheter obstruction.

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